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1	Simulation of Gear Dynamics by Circuit Theory Methods
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### 6 Abstract

A methodology of study of gear dynamics with the aid of the circuit theory and linear graph
methods is presented. In terms of analogy of the force, electric tension is used for the
composition of equivalent electrical circuit, which immediately gives the equations of motion.
The application of the electrical analogy method for the automation of composition of
equation of motion and their analysis are considered for the classical example of a one-stage
gear transmission with flexible supports and coupling masses. This technique can be extended
to analyse the dynamic characteristics of more complex dynamic systems as a planetary

14 transmission with flexible supports.

15

16 Index terms— dynamic gear model, equivalent electric circuit, linear graph, node equations, automation of 17 composition of motion equations.

## 18 1 Introduction

he classical method of Lagrange equation (Genkin and Grinkevich, 1961) and the methods of dynamic stiffness 19 or admittance (Airapetov et al. 1975) are used in studies of gear dynamics. These methods are cumbersome and 20 laborious. In order to automate composition of equations of motion, the bond graph (Karnopp and ??osenberg, 21 1972) and some other methods are also used. However, the use of these methods presupposes presentation of a 22 dynamic model as a system with concentrated parameters. Also, the bond graph of a relatively simple model, 23 as, for example, a planetary gear transmission, is very cumbersome (Allen, 1979). On the other hand, electrical 24 25 analogy method (Skudrzyk, 1968) and linear graph method (Mason and Zimmermann, 1960) allow us to model dynamic systems with both concentrated and distributed parameters (Podzharov, 1983, 1987, Sasa et al., 2004, 26 Wojnarowski, 2006, Kalous, 2009). 27 In this paper a methodology of study of gear dynamics with the aid of the electric circuit theory methods is 28

In this paper a methodology of study of gear dynamics with the aid of the electric circuit theory methods is presented. The analogy between the force and electric tension is used to compose equivalent electric circuits for dynamic gear systems. Nomenclature i J -moment of inertia of the mass i m , i k -torsional stiffness 3 2 , S S C C -support stiffnesses, 3 C -stiffness of tooth engagement, ) ( ), ( 2 1 t T t T -variable torsion moments, i µ -moment

of inertia reduced to the mass moving on the line of action, i C -torsion stiffness reduced to linear stiffness on the line of action, bi r -base radius of a gear, ) (t F itorsional moment reduced to a force applied in the line of action of gear engagement, ) (t S i -kinematic error in the gear engagement, ij y -element of the matrix of mechanical conductance, Y -matrix of mechanical conductance, Z -mechanical impedance, ij T -transmission in a graph between the points i and j, f -frequency, ?? = ??1.

## 38 **2** II.

# <sup>39</sup> 3 Modelling One-Stage Gear Transmission

We shall now consider the use of electrical analogy for the automation of composition of equations of motion and their analysis in the example of one-stage gear transmission with flexible supports and coupled masses. The mechanical model of the transmission is shown in Fig. ?? and in Fig. ??, presenting the equivalent electrical

#### MODELLING ONE-STAGE GEAR TRANSMISSION 3

circuit. Here, the parameters of the torsion system reduced to the parameters of a linear system are determined 43 as follows:2 / bi i i r J =  $\mu$ , 2 / bi i ki r k C = , bi i i r t T t F / ) ( ) ( = (1) 44

This system has 10 independent elements and 11 resonances and antiresonances (Skudrzyk, 1968), including 45 zero and infinite frequencies. When the circuit is excited by variable tensions (forces) Let us consider stationary 46 vibrations and assume that the gear is a linear dynamic system. Then, the solution of this system with periodic 47 force or kinematic excitation can be found as a sum of harmonics. In this case the equations of motion can be 48

composed as Kirchhoff equations of the equivalent electric circuit.) (1 t F and ) (2 t F the contours 49

The total conductance between the points a and b will be equal to the sum of the conductances of parallel 50 branches (Skudrzyk, 1968)., 4 3 2 1 0 Y Y Y Y Y Y Y ab + + + + = (2) where , / 2 0 C j Y ? = () , / 1 2 2 2 ? 51 52 3? + = ??jCmjYS1313441???????????????????????????????(3)53 The input impedance between points a and b can be found as inverse of abY ab ab Y Z / 1 = . (4) 54

Substituting the equations (3) in the equation (??) and (4) and making ab Z equal zero, we can find a 55 frequency characteristic of ab Z, which is shown in Fig. 3 for the transmission with the following parameters: 56 It was calculated neglecting the damping in the system and, according to the Foster theorem (Skudrzyk, 1968), 57 58 it has a monotonous character. We can find from the curve that the poles are at frequencies 55 Hz, 209 Hz, 300 59 Hz, 794 Hz and 5200 Hz. The zeros are at the frequencies 115 Hz, 259.5 Hz, 408 Hz and 859.5 Hz.

The poles pi f and zeros sj f can also be found approximately from the concepts of parallel and successive 60 resonances of resonances of tensions and currents: Hz C f p 50 2 1 4 1 1 =  $\mu$ ?, (5) Hz m C C f S p 232 2 1 3 3 61  $2 2 2 = + + = \mu$ ?, Hz m C f S p 336 2 1 2 1 2 1 3 = + + =  $\mu \mu$ ?, (7)Hz m C C f S p 778 2 1 2 23 2 1 4 = + 62  $+ = \mu ?$ ,(8) 63

() Hz m m C f p 5162 2 1 1 3 1 2 1 3 1 2 2 5 = + + + = ? ? ? ?  $\mu \mu$ ? (9) Hz C f S 104 2 1 3 3 1 = =  $\mu$ ? 64 (10)Hz m C f S S 5 . 259 2 1 3 3 2 = = ? , (11)Hz C f S 397 2 1 2 1 3 = =  $\mu$  ? , (12)Hz m C f S S 5 . 859 2 1 65  $2\ 2\ 4 = = ?$ , (13) 66

As we see, the approximate and exact frequencies are very similar to each other. Therefore, the formulas (5) 67 -(13) can be used for the analysis of resonances. 68

Now, let us use the linear graph method (Mason and Zimmermann, 1960) to obtain a general form for this 69 model. A linear graph for the circuit in Fig. ?? is constructed in Fig. ??. This graph illustrates the relations 70 between forces i F (upper nodes) and velocities i v 71

72 (lower nodes). The lines between them are transmissions, which in this case are mechanical admittances i y or mechanical impedances i z. Thus, in this graph we use relationsi i i F y v? = and i i i v z F? = (14)73

In order to simplify the graph, the number of nodes can be reduced retaining only the nodes which we need 74 to determine. Further simplification of the graph can be implemented by adding parallel transmissions and 75 excluding the nodes, which we do not need to determine, by splitting them. Hence, splitting the nodes i F  $\mu$ , Ci 76 v,  $i v \mu$  and excluding the loops i l, we can obtain the transformed graph presented in Fig. ??. 77

Here,) /( 1 i i j y  $\mu$  ?  $\mu$  = , i Ci C j y / ? = , )) /( /( 1 ?  $\mu$  ? j C j y i i Si + = . (15) 78

The transmissions of this graph can be determined by the following formulas ()1679

Where 1 2 1 1 / ) ( C y y y l µ µ + ? = , 3 4 3 3 / ) ( C y y y l µ µ + ? = , 2 3 2 3 2 2 / ) ( C S S y y y y l 80  $+++? = \mu \mu .$  (17) 81

This graph can also be described by the equations determining the nodes in relation to adjacent nodes and 82 transmissions that link them:) ( ) ( ) ( 4 33 3 2 32 2 22 3 23 2 1 21 1 11 2 12 1 t F T F F T t S T F T F F T t 83 F T F T F C C C C C C C = +? =? +? =?? (18) 84

Substituting (16) and (17) in (??8) and multiplying each of the itch equation (??8 t F F y F y t S F y F y 85 FytFyFyFyCCCCCCC+ +? =? +? =? ? $\mu(19)$ PFYC = ×,(21) 86

Here, the matrix of the mechanical conductance Y is symmetrical; each itch diagonal element is positive and 87 equal to the sum of the input mechanical conductances of the elements, which enter the itch node. Each non-88 diagonal element is negative and equal to the transition conductance of the elements, which locate between itch 89 and j-th nodes. This type of matrix is known in the circuit theory as matrix of node equations (Karni, 1966). 90 The excitation term in the right-hand part of the i-th equation is equal to the velocity of cinematic error in the 91 gear engagement, in the case of cinematic excitation. In the case of force excitation, it equals to the product of 92 the exciting force and the transition conductance between the point of application of the force and the itch node. 93 Therefore, it is not necessary to compose equivalent electric circuits and graphs. Instead, following the rules 94

explained above, we can directly compose the matrix of mechanical conductance and the vector of the right-hand 95 part of the equations. 96

The above formulated rules can be extended to more complicated systems and systems with distributed 97 parameters. 98

A dynamic calculation of the gear with the parameters mentioned above was implemented using the equations 99 (15) -(20). The damping in elastic elements was considered by introducing complex stiffness (Skudrzyk, 1968).) 100

1 ( i i i j C C ? + = , (22)101

Where i? -loss factor in i-th elastic element. 5) -(??). The measurement of noise and vibration of this gear 102 shows that it has high levels at frequency 5200 Hz. 103

# <sup>104</sup> 4 The results of calculation of dynamical forces

## 105 **5 III.**

## 106 6 Conclusions

a) The use of electric analogy allows us to avoid the derivation of equations of motion and to make a frequency
 analysis without solving the equations.

b) Equations analogous to node equations known in the circuit theory can be used in the gear dynamics forthe systems with concentrated and distributed parameters.

c) There is no need to compose any electric circuit and linear graph to obtain linear equations of motion of a

1 ? C 3 1 ? V µ1 V µ2 V µ3 V µ4 V c1 V c3 V 1 V 2 V c2 S t ? 2 ( )  $^{-1}$ 

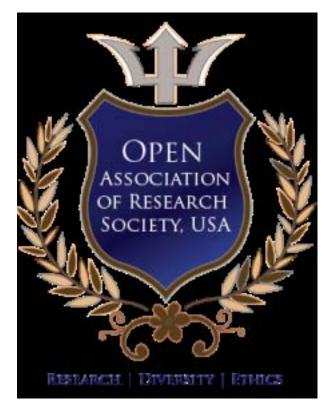


Figure 1:

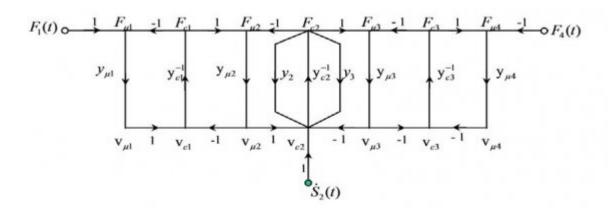


Figure 2: Where

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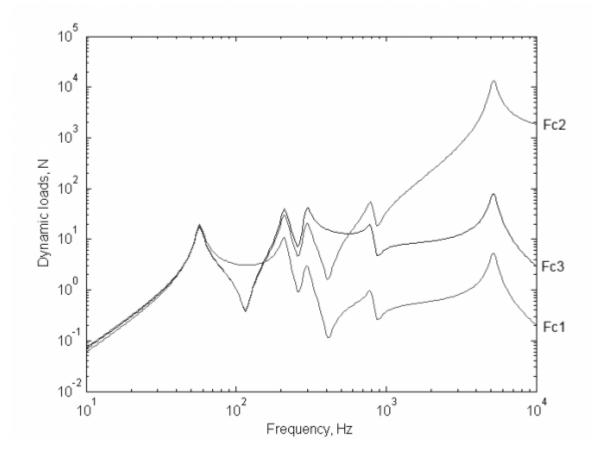


Figure 3:

1

No. Mechanical System Electrical System			
1	Force	Tension	
2	Speed	Electric current	
3	Displacement	Electric charge	
4	Mass	Inductance	
5	Flexibility	Capacitance	
6	Absorber	Electrical resistance	

Figure 4: Table 1 :

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